



Analysis of Inductive WPT

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Outline

- » Introduction
- » Efficiency
- » Power balance
- » Input impedance
- » Conclusion

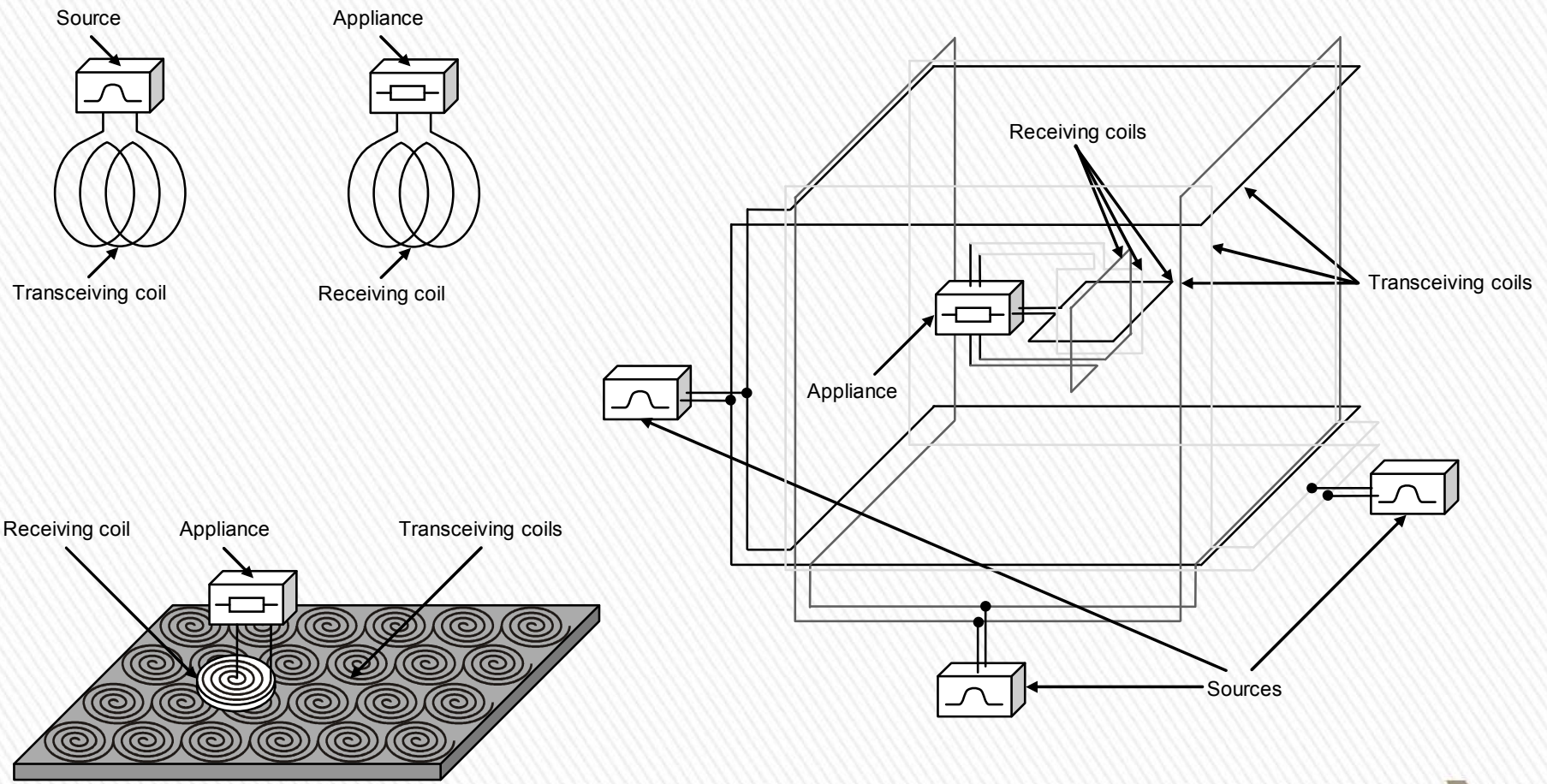


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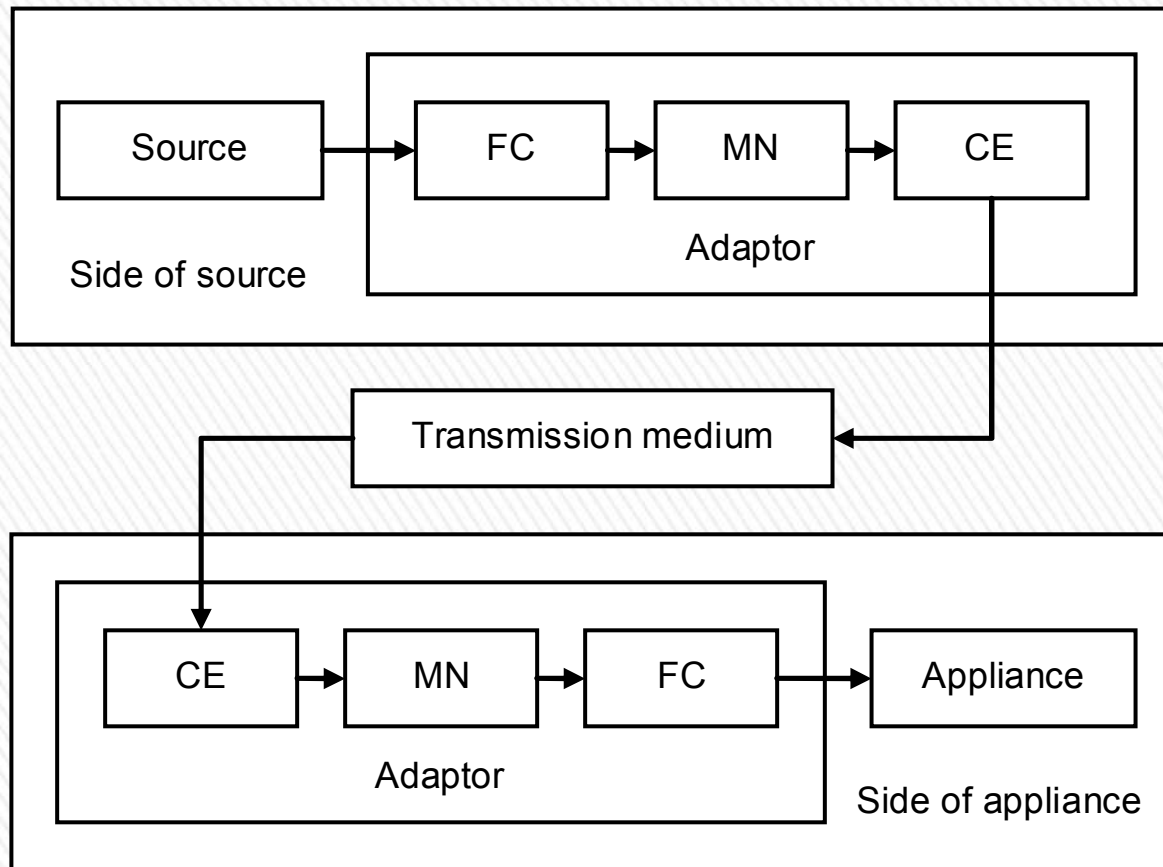
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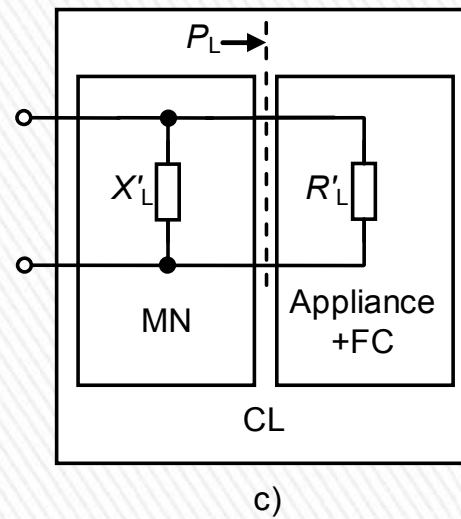
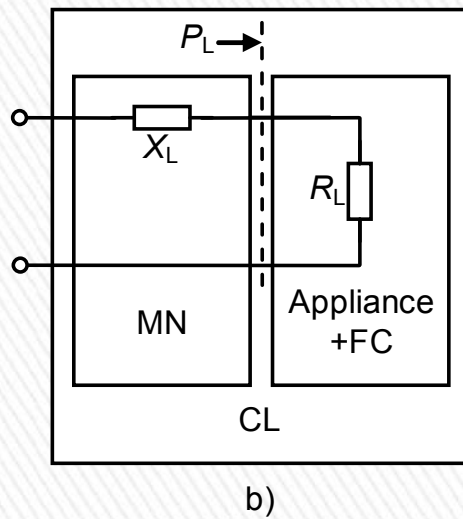
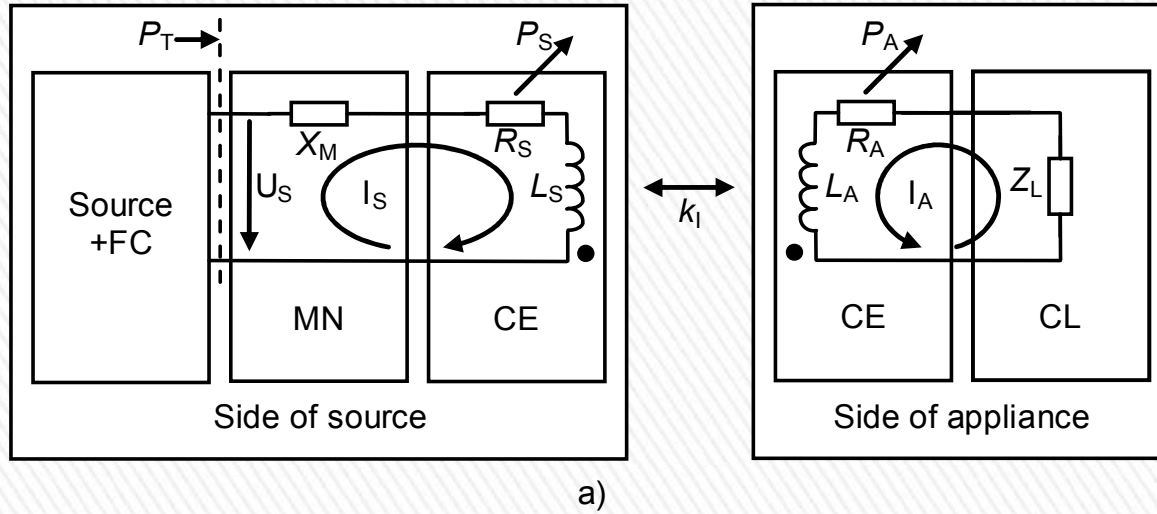
Introduction



General WPT Chain



Circuit Model



Outline

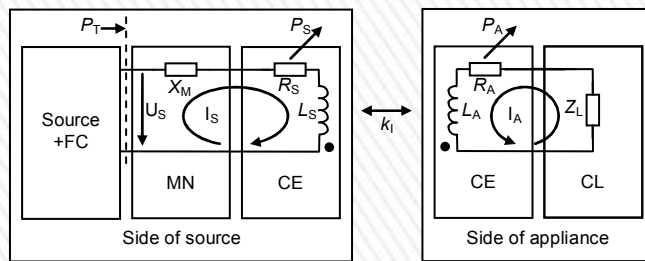
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- » **Efficiency**
- » Power balance
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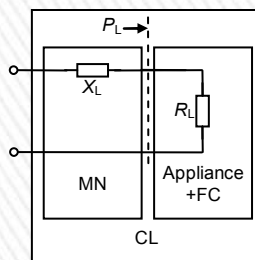
Efficiency Definition

» Main losses are caused by resistances of transceiving and receiving coils.

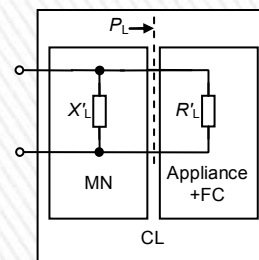
$$\eta_1 = \frac{P_L}{P_T} = \frac{P_L}{P_A + P_L + P_S}$$



a)



b)



c)

Efficiency Derivation (1)

$$\gg \eta_i = \frac{R_L X_K^2}{\left((R_A + R_L)^2 + (X_A + X_L)^2 \right) R_S + (R_A + R_L) X_K^2}$$

$$X_A = \omega L_A, X_S = \omega L_S, X_K = \omega k_I \sqrt{L_A L_S}$$

$$\frac{X_K}{\sqrt{X_A X_S}} = k_I, \frac{X_A}{R_A} = Q_A, \frac{X_L}{R_A} = -Q'_A, \frac{X_A}{R_L} = Q_L, \frac{X_S}{R_S} = Q_S$$

$$k_I^2 Q_A Q_S \frac{Q_A}{Q_L}$$

$$\gg \eta_i = \frac{k_I^2 Q_A Q_S \frac{Q_A}{Q_L}}{\left(1 + \frac{Q_A}{Q_L} \right)^2 + (Q_A - Q'_A)^2 + k_I^2 Q_A Q_S \left(1 + \frac{Q_A}{Q_L} \right)}$$



Efficiency Derivation (2)

$$\gg \kappa = k_I \sqrt{Q_A Q_S}, \quad \rho = \frac{Q_A}{Q_L}, \quad \xi = Q_A - Q'_A$$

$$\eta_I = \frac{\kappa^2 \rho}{(1 + \rho)^2 + \xi^2 + \kappa^2 (1 + \rho)}$$

$$\gg R, L, C, \omega, (X) \rightarrow k_I, Q_A, Q'_A, Q_L, Q_S \rightarrow \kappa, \rho, \xi \rightarrow$$

$$\max(\eta_I(\rho, \xi)) \rightarrow \rho = \sqrt{1 + \kappa^2}, \quad \xi = 0 \rightarrow$$

$$\kappa, \rho, \xi \rightarrow k_I, Q_A, Q'_A, Q_L, Q_S \rightarrow R, L, C, \omega, (X)$$

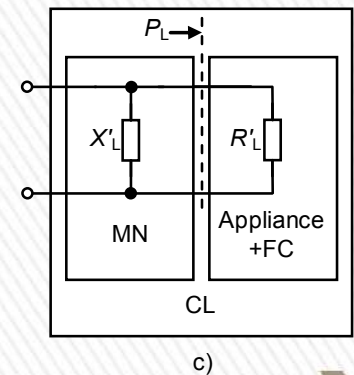
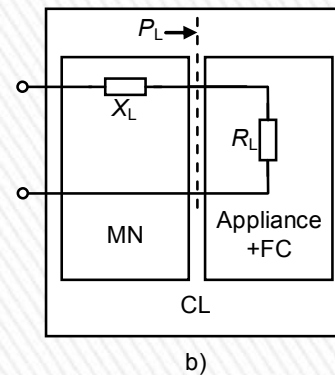
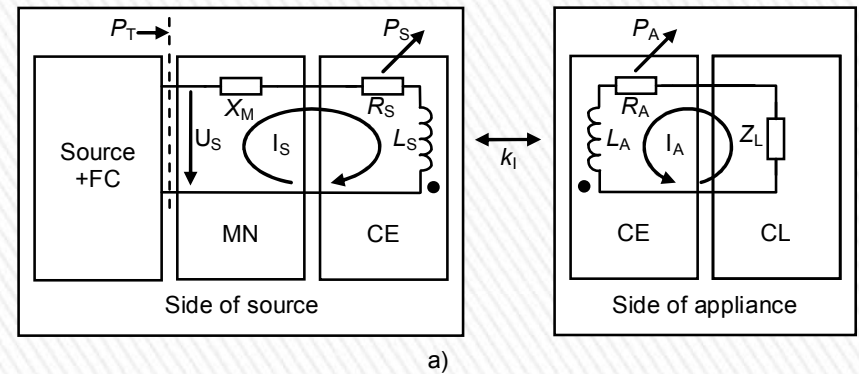
Maximal Efficiency, Optimal Conditions

$$\gg R_L = \sqrt{R_A^2 + \omega^2 k_I^2 L_A L_S} \frac{R_A}{R_S}$$

$$X_L = -\omega L_A$$

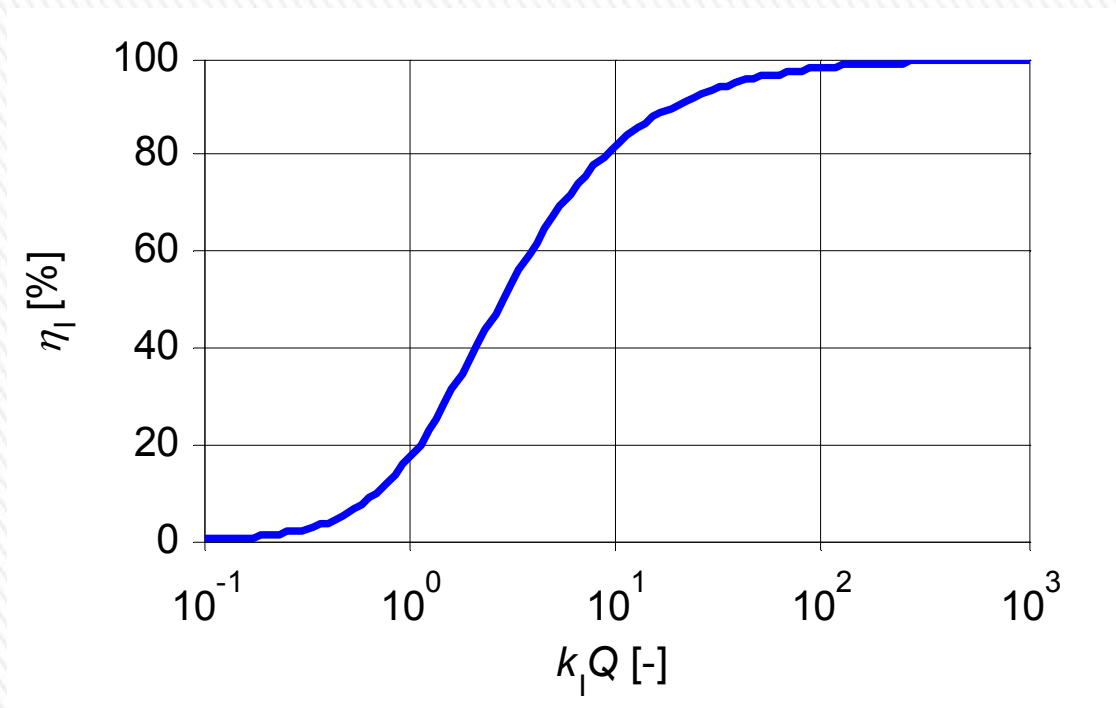
$$\gg \eta_I = \frac{(k_I Q)^2}{\left(1 + \sqrt{1 + (k_I Q)^2}\right)^2}$$

$$Q = \sqrt{Q_A Q_S}$$



Maximal Efficiency

$$\gg \eta_1 = \frac{(k_1 Q)^2}{\left(1 + \sqrt{1 + (k_1 Q)^2}\right)^2}$$

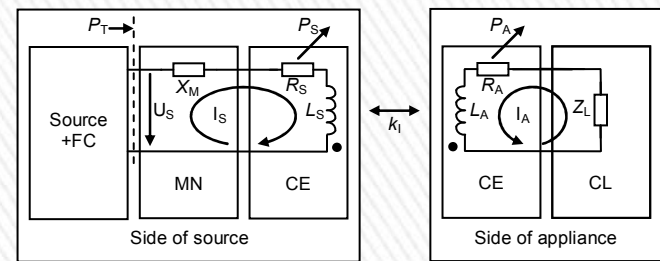


Max. eff., Opt. Con. – Dual Case

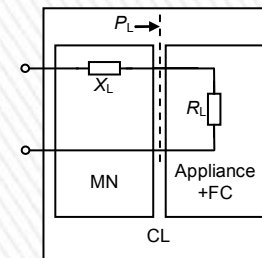
$$\gg R_L + jX_L = \frac{R'_L X'_L{}^2}{R'_L{}^2 + X'_L{}^2} + j \frac{R'_L{}^2 X'_L}{R'_L{}^2 + X'_L{}^2}$$

$$\gg R'_L = \frac{\omega^2 L_A^2}{\sqrt{R_A^2 + \omega^2 k_I^2 L_A L_S} \frac{R_A}{R_S}} + \sqrt{R_A^2 + \omega^2 k_I^2 L_A L_S} \frac{R_A}{R_S}$$

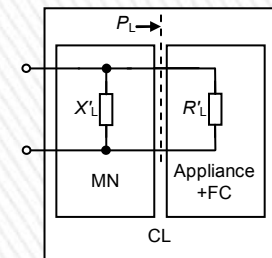
$$X'_L = -\omega L_A - \frac{R_A^2 + \omega^2 k_M^2 L_A L_S}{\omega L_A} \frac{R_A}{R_S}$$



a)



b)



c)

Outline

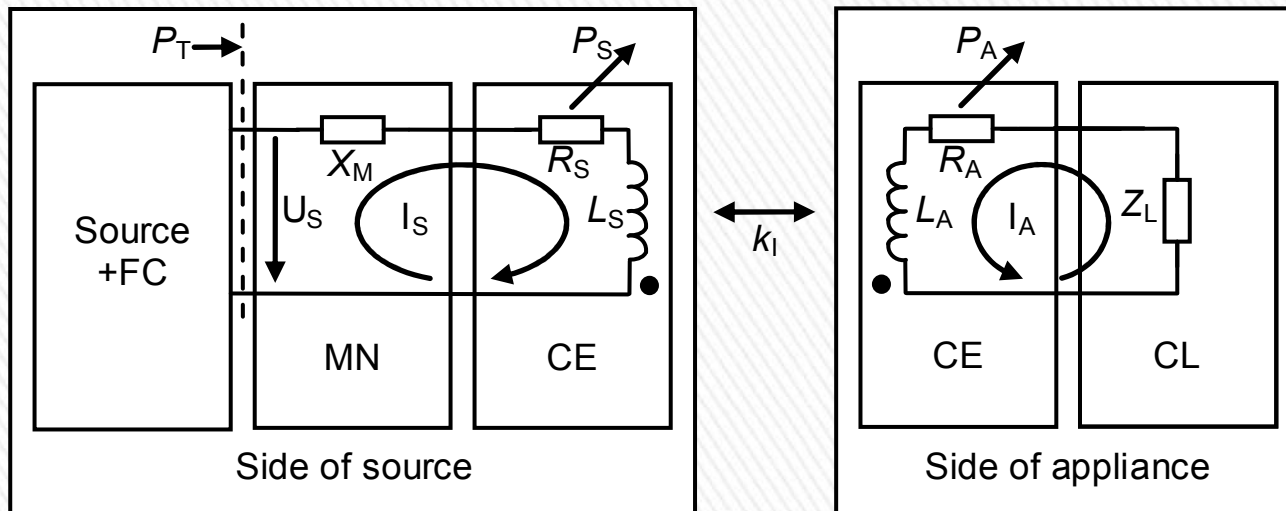
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Maximal Appliance Power, Optimal Condition

» $\max(P_L(X_M))$

» $X_M = -\omega L_S$

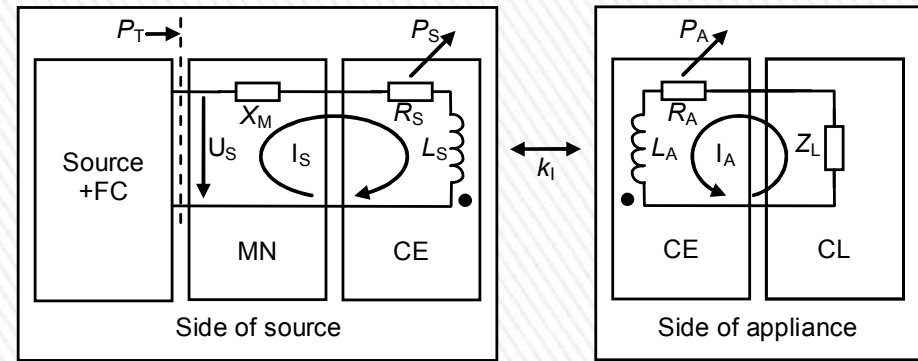


Power Balance (1)

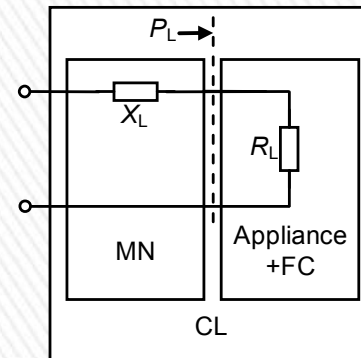
$$\gg R_L = \sqrt{R_A^2 + \omega^2 k_I^2 L_A L_S} \frac{R_A}{R_S}$$

$$\gg X_L = -\omega L_A$$

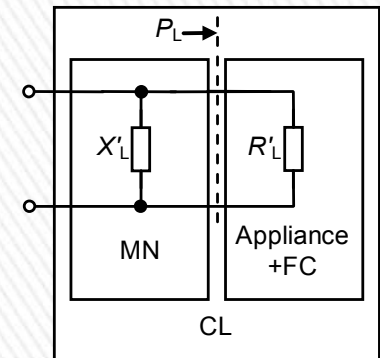
$$\gg X_M = -\omega L_S$$



a)



b)



c)



Power Balance (2)

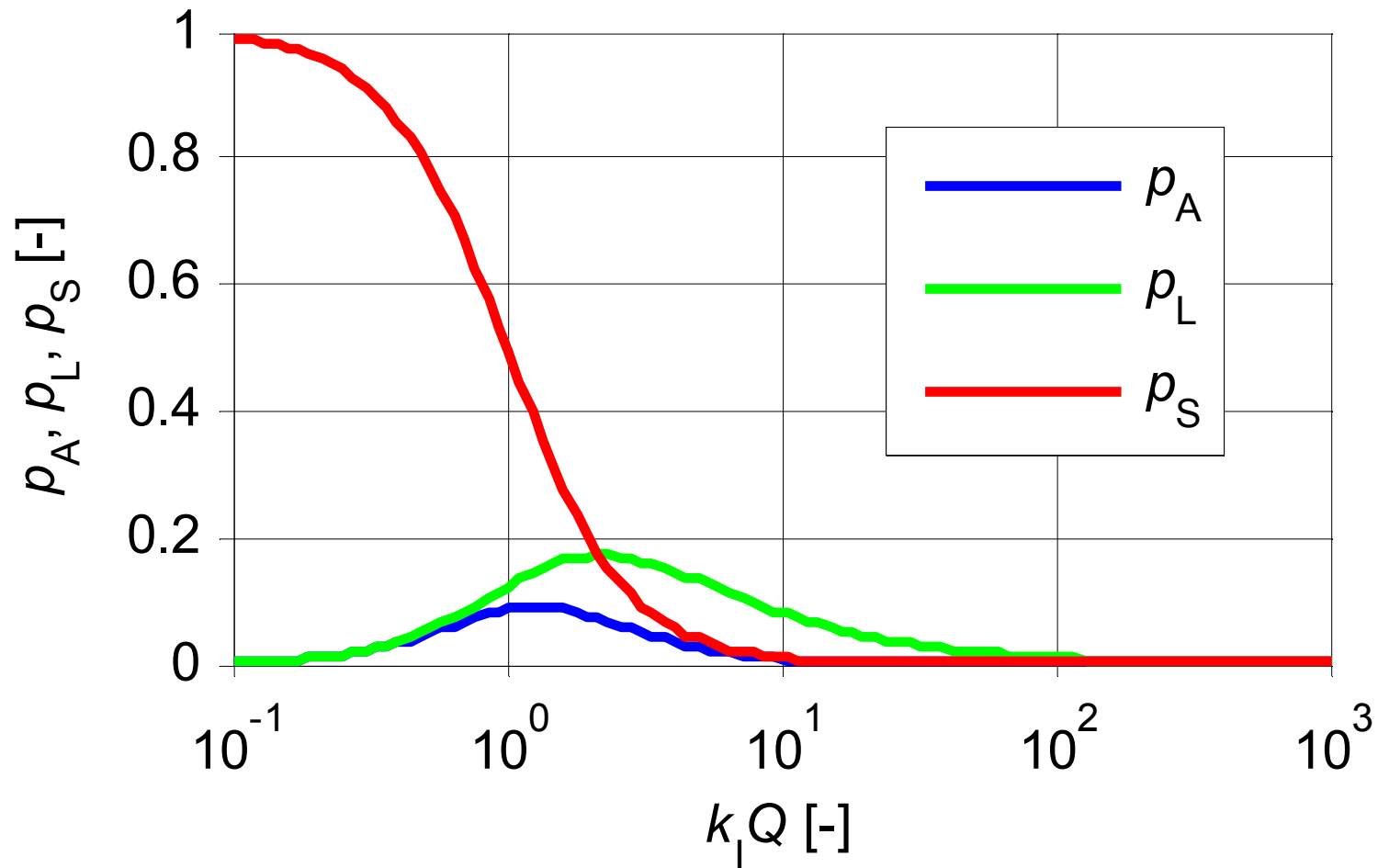
$$\gg P_A = \frac{(k_I Q)^2}{\underbrace{\left(1 + (k_I Q)^2 + \sqrt{1 + (k_I Q)^2}\right)^2}_{\rho_A}} \frac{|U_S|^2}{R_S}$$

$$\gg P_L = \frac{(k_I Q)^2 \sqrt{1 + (k_I Q)^2}}{\underbrace{\left(1 + (k_I Q)^2 + \sqrt{1 + (k_I Q)^2}\right)^2}_{\rho_L}} \frac{|U_S|^2}{R_S}$$

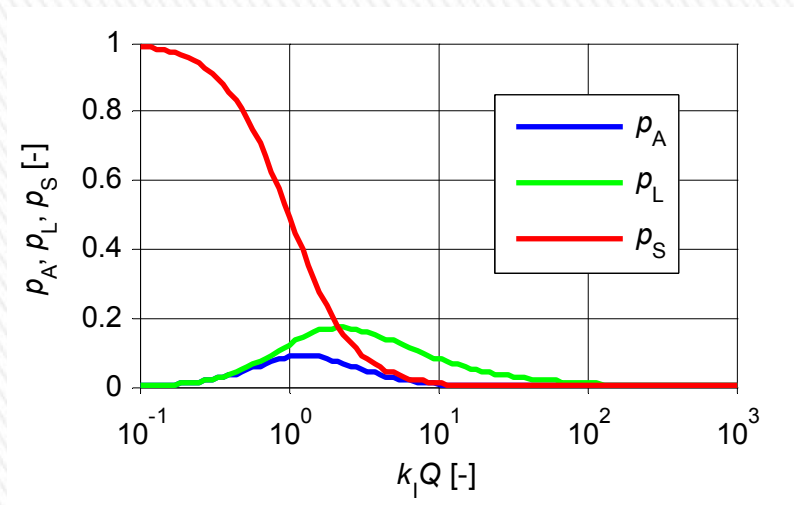
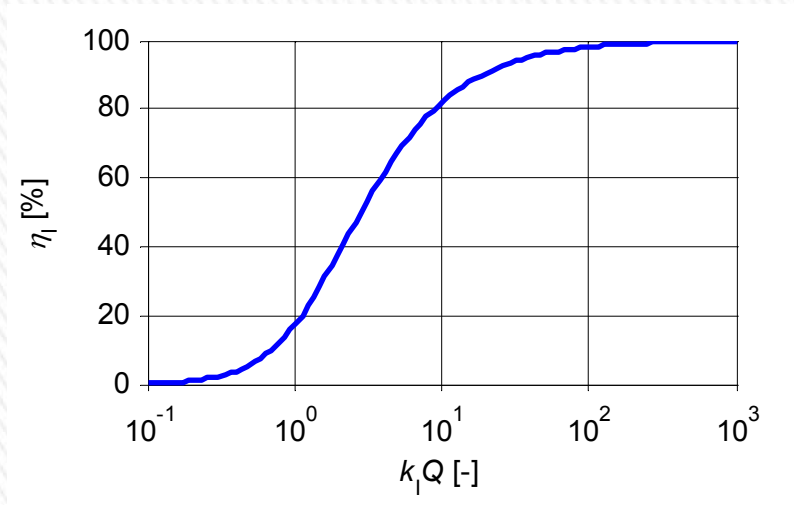
$$\gg P_S = \frac{\left(1 + \sqrt{1 + (k_I Q)^2}\right)^2}{\underbrace{\left(1 + (k_I Q)^2 + \sqrt{1 + (k_I Q)^2}\right)^2}_{\rho_S}} \frac{|U_S|^2}{R_S}$$



Power Balance (3)



Efficiency and Power Balance



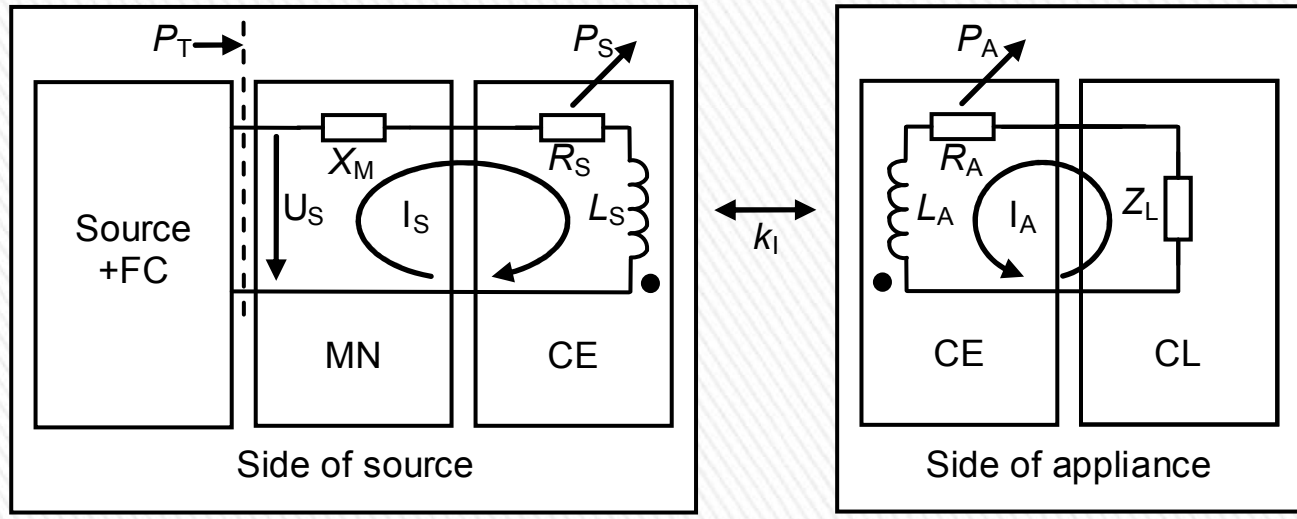
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Input Impedance (1)

$$\gg Z_S = \frac{U_S}{I_S}$$



Input Impedance (2)

$$\begin{aligned} \gg Z_S = \frac{U_S}{I_S} = R_S + \frac{(R_A + R_L) X_K^2}{(R_A + R_L)^2 + (X_A + X_L)^2} + \\ + j \left(X_M + X_S - \frac{(X_A + X_L) X_K^2}{(R_A + R_L)^2 + (X_A + X_L)^2} \right) \end{aligned}$$

» By optimal conditions for efficiency and appliance power:

$$Z_S = R_S + \frac{\omega^2 k_I^2 L_A L_S}{R_A + \sqrt{R_A^2 + \omega^2 k_I^2 L_A L_S \frac{R_A}{R_S}}}$$

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Conclusion

- » The general circuit model for WPT by electromagnetic induction was shown.
- » The efficiency was mentioned.
- » The power balance in the terms of normalized powers was derived.
- » The input impedance was presented.

